

# Near-Optimal Training-Based Estimation of Frequency Offset and Channel Response in OFDM with Phase Noise

Darryl Dexu Lin, Ryan Pacheco, Teng Joon Lim and Dimitrios Hatzinakos

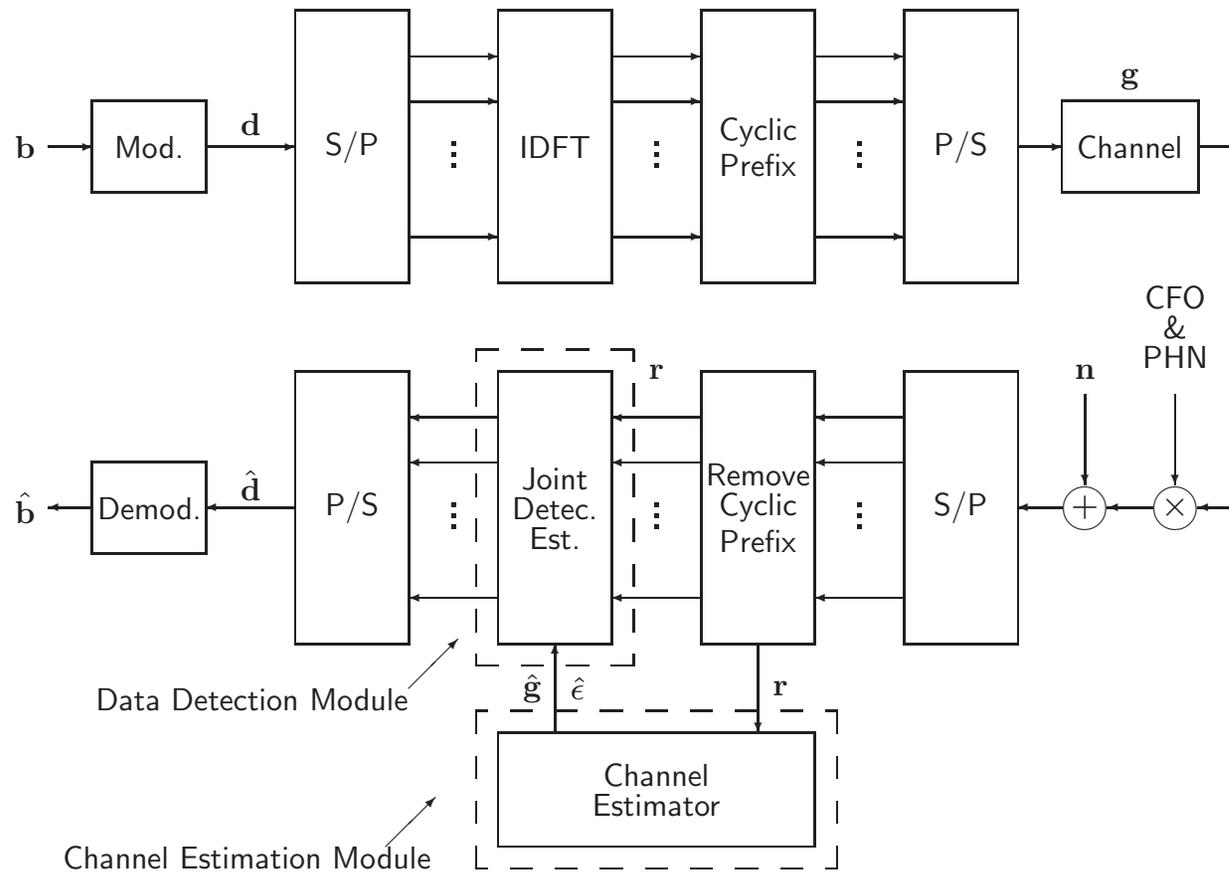
*Presented by Darryl Dexu Lin at ICC'06*

*June 12, 2006*

- **Problem description and signal model.**
- Prior statistics of phase noise.
- The optimal joint CFO/PHN/CIR estimator.
- The near-optimal joint CFO/PHN/CIR estimator.
- Complexity reduction using conjugate gradient method.
- Simulation results.

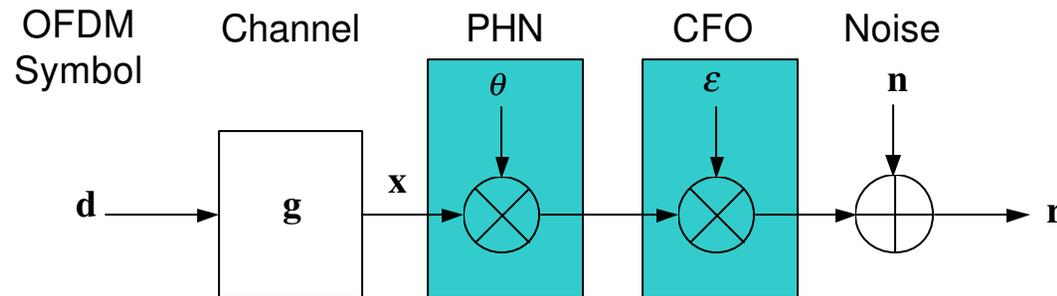
# I. Problem Description

- The role of channel estimation in OFDM receiver design:

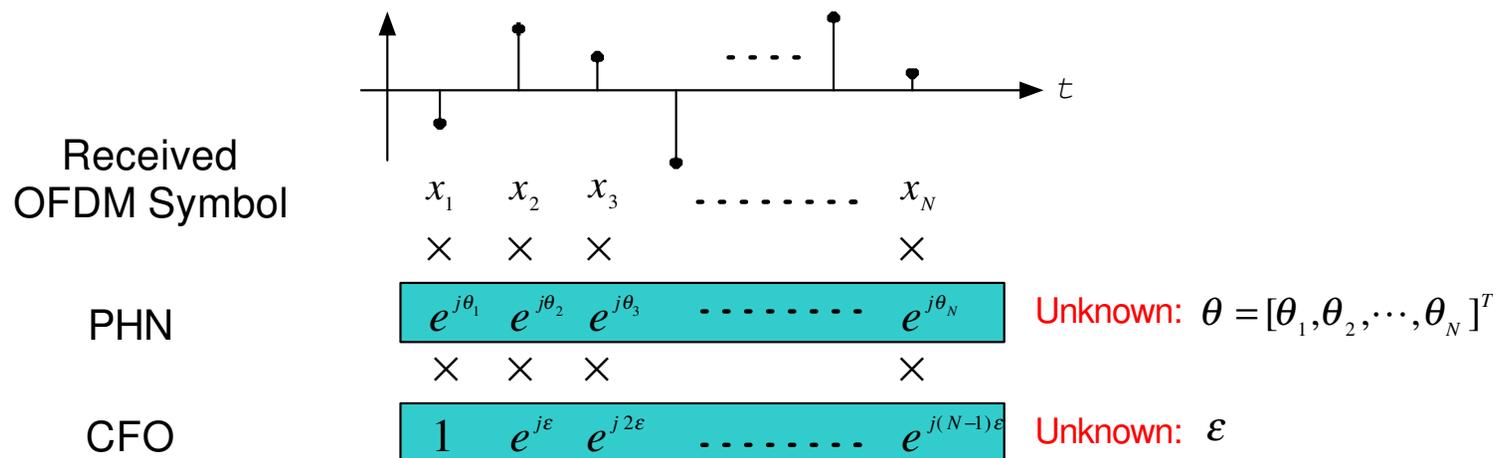


# I. Problem Description

- OFDM channel with PHN (Phase Noise) and CFO (Carrier Frequency Offset):



- Distortions caused by PHN and CFO:



# I. Signal Model in Matrix Form

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- Complex baseband received signal in one OFDM symbol interval:

$$\mathbf{r} = \mathbf{E}\mathbf{P}\mathbf{G}\mathbf{F}^H\mathbf{d} + \mathbf{n}, \quad (1)$$

- $\mathbf{r} \in \mathbb{C}^{N \times 1}$ : received OFDM symbol with cyclic prefix removed;
  - $\mathbf{E} = \text{diag}([1, e^{j2\pi\epsilon/N}, \dots, e^{j2\pi(N-1)\epsilon/N}]^T)$ : CFO matrix;
  - $\mathbf{P} = \text{diag}([e^{j\theta_0}, \dots, e^{j\theta_{N-1}}]^T)$ : PHN matrix;
  - $\mathbf{G}$ : channel circular convolution matrix, formed by CIR  $\mathbf{g}$ ;
  - $\mathbf{F} \in \mathbb{C}^{N \times N}$ : DFT matrix;
  - $\mathbf{d} \in \mathbb{C}^{N \times 1}$ : vector of constant-modulus training symbols;
  - $\mathbf{n} \in \mathbb{C}^{N \times 1}$ : complex white Gaussian noise with variance  $\sigma^2$  per dimension.
- The objective is to, based on received  $\mathbf{r}$ , estimate three unknowns:
    - (1) $\epsilon$ , (2) $\boldsymbol{\theta} = [\theta_0, \dots, \theta_{N-1}]^T$ , (3) $\mathbf{g} = [g_0, \dots, g_{L-1}]^T$ .

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- Problem description and signal model.
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## II. Prior Statistics of Phase Noise

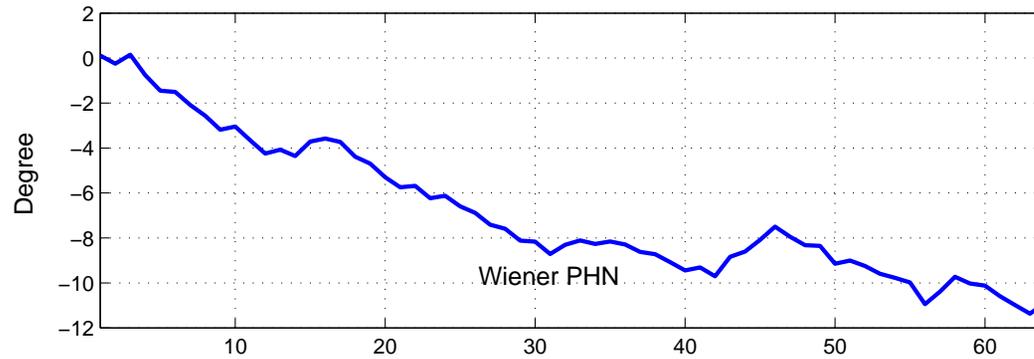
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- Two different models of PHN are available:
  - For free-running oscillator at the receiver, we assume a non-stationary Gaussian process, called *Wiener PHN*.
  - For oscillator controlled by a phase-locked loop (PLL), we assume a zero-mean coloured Gaussian process, called *Gaussian PHN*.
- The prior statistics of both types of PHN can be modeled as a multivariate Gaussian distribution:

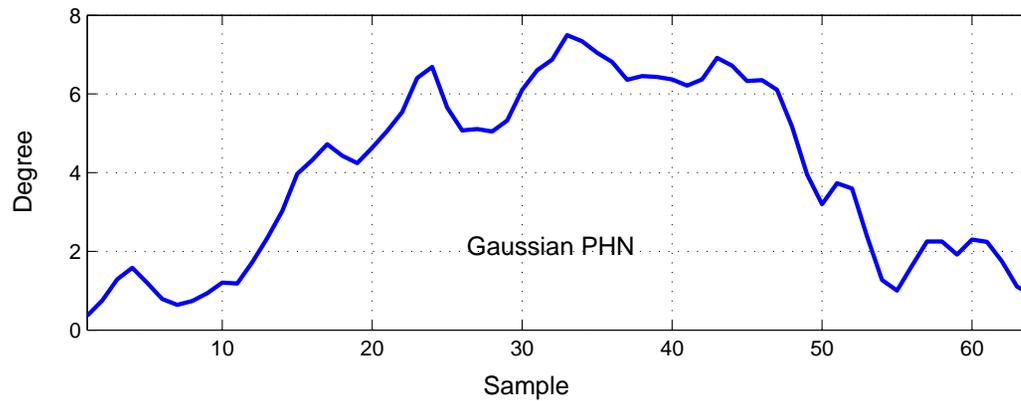
$$p(\boldsymbol{\theta}) = \mathcal{N}(\mathbf{0}, \boldsymbol{\Phi}), \quad (2)$$

where the covariance matrix  $\boldsymbol{\Phi}$  can be determined from the power spectral density (PSD) of the VCO output.

## II. Prior Statistics of Phase Noise



$$\Phi = \alpha_{\phi}^2 \begin{bmatrix} 1 & 1 & \cdots & 1 \\ 1 & 2 & \cdots & 2 \\ \vdots & \vdots & \ddots & \vdots \\ 1 & 2 & \cdots & N \end{bmatrix}$$



$$\Phi = \begin{bmatrix} \phi_0 & \phi_1 & \cdots & \phi_{N-1} \\ \phi_1 & \phi_0 & \cdots & \vdots \\ \vdots & \cdots & \ddots & \phi_1 \\ \phi_{N-1} & \cdots & \phi_1 & \phi_0 \end{bmatrix}$$

PHN Sample

Covariance Matrix

- 
- Problem description and signal model.
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### III. The Complete Likelihood Function

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- The optimal estimator requires joint estimation of three unknowns,  $\epsilon$ ,  $\boldsymbol{\theta}$  and  $\mathbf{g}$ , in (3), where  $\mathbf{F} = [\mathbf{W}|\mathbf{V}]$  and  $\mathbf{g}$  is the CIR.

$$\mathbf{r} = \mathbf{E}\mathbf{P}\mathbf{F}^H\mathbf{D}\mathbf{W}\mathbf{g} + \mathbf{n}. \quad (3)$$

- We first write the “complete likelihood function”

$$p(\mathbf{r}, \epsilon, \boldsymbol{\theta}, \mathbf{g}) = p(\mathbf{r}|\epsilon, \boldsymbol{\theta}, \mathbf{g})p(\epsilon)p(\boldsymbol{\theta})p(\mathbf{g}), \quad (4)$$

which is proportional to the *a posteriori* distribution of the unknowns,  $p(\epsilon, \boldsymbol{\theta}, \mathbf{g}|\mathbf{r})$ .

- Since we assume no prior knowledge of  $\epsilon$  and  $\mathbf{g}$ ,  $p(\epsilon)$  and  $p(\mathbf{g})$  are constants and can be omitted. The prior of  $\boldsymbol{\theta}$  is available, which is  $p(\boldsymbol{\theta}) = \mathcal{N}(\mathbf{0}, \boldsymbol{\Phi})$  as discussed before.

### III. Target of Optimization

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- Taking the logarithm, the “complete negative log-likelihood function” can be written as

$$\begin{aligned}\mathcal{L}(\epsilon, \boldsymbol{\theta}, \mathbf{g}) &= -\log p(\mathbf{r}|\epsilon, \boldsymbol{\theta}, \mathbf{g}) - \log p(\boldsymbol{\theta}) \\ &= \frac{1}{2\sigma^2}(\mathbf{r} - \mathbf{E}\mathbf{P}\mathbf{F}^H\mathbf{D}\mathbf{W}\mathbf{g})^H(\mathbf{r} - \mathbf{E}\mathbf{P}\mathbf{F}^H\mathbf{D}\mathbf{W}\mathbf{g}) + \frac{1}{2}\boldsymbol{\theta}^T\boldsymbol{\Phi}^{-1}\boldsymbol{\theta}.\end{aligned}\tag{5}$$

- The objective is to find the jointly optimal estimates

$$(\hat{\epsilon}, \hat{\boldsymbol{\theta}}, \hat{\mathbf{g}}) = \arg \min_{\epsilon, \boldsymbol{\theta}, \mathbf{g}} \mathcal{L}(\epsilon, \boldsymbol{\theta}, \mathbf{g}).\tag{6}$$

- The estimator proposed here is “optimal” in the sense of maximizing the “complete likelihood function”. It can be derived in 3 optimization steps.

### III. The Optimal Estimator: CIR and PHN Estimation

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1. *CIR Estimation*: Solve  $\partial\mathcal{L}(\epsilon, \boldsymbol{\theta}, \mathbf{g})/\partial\mathbf{g}^* = \mathbf{0}$ , we obtain

$$\hat{\mathbf{g}} = (2\rho^2)^{-1}\mathbf{W}^H\mathbf{D}^H\mathbf{F}\mathbf{P}^H\mathbf{E}^H\mathbf{r}. \quad (7)$$

– Substituting  $\mathbf{g} = \hat{\mathbf{g}}$  back into  $\mathcal{L}(\epsilon, \boldsymbol{\theta}, \mathbf{g})$  produces  $\mathcal{L}(\epsilon, \boldsymbol{\theta})$ .

2. *PHN Estimation*: Solve  $\partial\mathcal{L}(\epsilon, \boldsymbol{\theta})/\partial\boldsymbol{\theta} = \mathbf{0}$ , we obtain

$$\hat{\boldsymbol{\theta}} = [\text{Re}(\mathbf{E}\mathbf{C}\mathbf{C}^H\mathbf{E}^H) + 2\sigma^2\rho^2\boldsymbol{\Phi}^{-1}]^{-1}\text{Im}(\mathbf{E}\mathbf{C}\mathbf{C}^H\mathbf{E}^H)\mathbf{1}, \quad (8)$$

where  $\mathbf{C} = \mathbf{R}^H\mathbf{F}^H\mathbf{D}\mathbf{V}$  and  $\mathbf{R} = \text{diag}(\mathbf{r})$ .

– Substituting  $\boldsymbol{\theta} = \hat{\boldsymbol{\theta}}$  back into  $\mathcal{L}(\epsilon, \boldsymbol{\theta})$  produces  $\mathcal{L}(\epsilon)$

### III. The Optimal Estimator: CFO Estimation

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3. *CFO Estimation*: To minimize  $\mathcal{L}(\epsilon)$ , we require searching over the range  $-0.5 < \epsilon < 0.5$ :

$$\hat{\epsilon} = \arg \min_{\epsilon} \mathbf{1}^T \mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H \mathbf{1} - \mathbf{1}^T \text{Im}(\mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H)^T \times [\text{Re}(\mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H) + 2\sigma^2 \rho^2 \mathbf{\Phi}^{-1}]^{-1} \text{Im}(\mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H) \mathbf{1}. \quad (9)$$

- Combining the 3 optimization steps leads to the complete *Joint CFO/PHN/CIR Estimation* (JCPCE) algorithm (Proc. WCNC'06):
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$$\begin{aligned} \text{Step 1: } \hat{\epsilon} &= \arg \min_{\epsilon} \mathbf{1}^T \mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H \mathbf{1} - \mathbf{1}^T \text{Im}(\mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H)^T \\ &\quad \times [\text{Re}(\mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H) + 2\sigma^2 \rho^2 \mathbf{\Phi}^{-1}]^{-1} \text{Im}(\mathbf{E} \mathbf{C} \mathbf{C}^H \mathbf{E}^H) \mathbf{1}; \\ \hat{\mathbf{E}} &= \text{diag}([1, e^{j2\pi\hat{\epsilon}/N}, \dots, e^{j2\pi(N-1)\hat{\epsilon}/N}]^T); \\ \text{Step 2: } \hat{\boldsymbol{\theta}} &= [\text{Re}(\hat{\mathbf{E}} \mathbf{C} \mathbf{C}^H \hat{\mathbf{E}}^H) + 2\sigma^2 \rho^2 \mathbf{\Phi}^{-1}]^{-1} \text{Im}(\hat{\mathbf{E}} \mathbf{C} \mathbf{C}^H \hat{\mathbf{E}}^H) \mathbf{1}; \\ \hat{\mathbf{P}} &= \text{diag}([e^{j\hat{\theta}_0}, \dots, e^{j\hat{\theta}_{N-1}}]^T); \\ \text{Step 3: } \hat{\mathbf{g}} &= (2\rho^2)^{-1} \mathbf{W}^H \mathbf{D}^H \mathbf{F} \hat{\mathbf{P}}^H \hat{\mathbf{E}}^H \mathbf{r}. \end{aligned}$$


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## IV. Preliminary: Moose's CFO Estimator

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- Assuming only CFO and no PHN, what is the simple way to estimate the CFO without searching over all possible values of  $\epsilon$ ?
- The Moose's method proposes to transmit a training sequence with two identical halves in the time domain, of length  $N/2$ . Thus, the received vector  $\mathbf{r} = [\mathbf{r}_1^T, \mathbf{r}_2^T]^T$ , where

$$\mathbf{r}_1 = \mathbf{x} + \mathbf{n}_1; \quad \mathbf{r}_2 = e^{j\pi\epsilon} \mathbf{x} + \mathbf{n}_2 \quad (10)$$

in which  $\mathbf{x} = \mathbf{E}\mathbf{G}\mathbf{F}^H \mathbf{d} \in \mathbb{C}^{\frac{N}{2} \times 1}$ .

- The maximum-likelihood estimate of  $\epsilon$  is

$$\hat{\epsilon} = \arg \max_{\epsilon} p(\mathbf{r}_1, \mathbf{r}_2 | \epsilon) = \arg \max_{\epsilon} p(\mathbf{r}_2 | \epsilon, \mathbf{r}_1) p(\mathbf{r}_1 | \epsilon). \quad (11)$$

## IV. Preliminary: Moose's CFO Estimator (Cont.)

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- If we assume the independence between  $\mathbf{r}_1$  and  $\epsilon$ , i.e.  $p(\mathbf{r}_1|\epsilon) = p(\mathbf{r}_1)$ , we simply have  $\hat{\epsilon} = \arg \max_{\epsilon} p(\mathbf{r}_2|\epsilon, \mathbf{r}_1)$ .
- Notice that  $\mathbf{r}_2 = e^{j\pi\epsilon}\mathbf{r}_1 - e^{j\pi\epsilon}\mathbf{n}_1 + \mathbf{n}_2 = e^{j\pi\epsilon}\mathbf{r}_1 + \mathbf{z}$ .
- Since  $\mathbf{n}_1, \mathbf{n}_2 \sim \mathcal{CN}(\mathbf{0}, 2\sigma^2\mathbf{I}) \Rightarrow \mathbf{z} \sim \mathcal{CN}(\mathbf{0}, 4\sigma^2\mathbf{I})$ , we have

$$-\log p(\mathbf{r}_2|\epsilon, \mathbf{r}_1) = \frac{1}{4\sigma^2}(\mathbf{r}_2 - e^{j\pi\epsilon}\mathbf{r}_1)^H(\mathbf{r}_2 - e^{j\pi\epsilon}\mathbf{r}_1). \quad (12)$$

- Thus  $\hat{\epsilon} = \arg \max_{\epsilon} p(\mathbf{r}_2|\epsilon, \mathbf{r}_1)$  leads to the near-ML estimate of  $\epsilon$

$$\hat{\epsilon} = \frac{1}{\pi} \angle \mathbf{r}_1^H \mathbf{r}_2. \quad (13)$$

## IV. Near Optimal Estimator: The New Approach

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- How do we make use of the convenience offered by repeating training sequence to efficiently estimate CIR and PHN together with CFO?
- We propose a two stage process:
  1. Estimate  $\hat{\epsilon} = \arg \max_{\epsilon} p(\mathbf{r}_1, \mathbf{r}_2 | \epsilon)$ , motivated by Moose's method.
  2. Estimate  $(\hat{\boldsymbol{\theta}}, \hat{\mathbf{g}}) = \arg \min_{\boldsymbol{\theta}, \mathbf{g}} \mathcal{L}(\hat{\epsilon}, \boldsymbol{\theta}, \mathbf{g})$ , similar to the optimal algorithm proposed earlier, except with  $\hat{\epsilon}$  already estimated, where

$$\begin{aligned} & \mathcal{L}(\epsilon, \boldsymbol{\theta}, \mathbf{g}) \\ = & -\log p(\mathbf{r} | \epsilon, \boldsymbol{\theta}, \mathbf{g}) - \log p(\boldsymbol{\theta}) \\ = & \frac{1}{2\sigma^2} (\mathbf{r} - \mathbf{E}\mathbf{P}\mathbf{F}^H \mathbf{D}\mathbf{W}\mathbf{g})^H (\mathbf{r} - \mathbf{E}\mathbf{P}\mathbf{F}^H \mathbf{D}\mathbf{W}\mathbf{g}) + \frac{1}{2} \boldsymbol{\theta}^T \boldsymbol{\Phi}^{-1} \boldsymbol{\theta}. \end{aligned} \tag{14}$$

## IV. Near Optimal Estimator: CFO Estimation

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- *CFO Estimation*: In the presence of PHN, CFO estimation using the Moose's method needs to be modified, since

$$\mathbf{r}_1 = \mathbf{P}_1 \mathbf{x} + \mathbf{n}_1; \quad \mathbf{r}_2 = e^{j\pi\epsilon} \mathbf{P}_2 \mathbf{x} + \mathbf{n}_2, \quad (15)$$

where  $\mathbf{P}_1$  and  $\mathbf{P}_2$  contain consecutive PHN sequences  $\theta_1$  and  $\theta_2$ .

- The optimal estimate  $\hat{\epsilon}$  is then

$$\arg \max_{\epsilon} p(\mathbf{r}_1, \mathbf{r}_2 | \epsilon) = \arg \max_{\epsilon} \int_{\theta_2} \int_{\theta_1} p(\mathbf{r}_1, \mathbf{r}_2, \theta_1, \theta_2 | \epsilon) d\theta_1 d\theta_2 \quad (16)$$

- Writing  $\mathbf{R}_1 = \text{diag}(\mathbf{r}_1)$  and  $\Phi_{\Delta}$  as the covariance matrix of  $\theta_2 - \theta_1$ , it can be shown that this is equivalent to maximizing

$$p(\mathbf{r}_2 | \mathbf{r}_1, \epsilon) = \mathcal{CN}(e^{j\pi\epsilon} \mathbf{r}_1, \mathbf{R}_1 \Phi_{\Delta} \mathbf{R}_1^H + 4\sigma^2 \mathbf{I}). \quad (17)$$

## IV. Near Optimal Estimator: CIR Estimation

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- Therefore,

$$\hat{\epsilon} = \frac{1}{\pi} \angle \mathbf{r}_1^H (\mathbf{R}_1 \Phi_{\Delta} \mathbf{R}_1^H + 4\sigma^2 \mathbf{I})^{-1} \mathbf{r}_2. \quad (18)$$

- Substituting  $\hat{\epsilon}$  into  $\mathcal{L}(\epsilon, \boldsymbol{\theta}, \mathbf{g})$ , we get

$$\mathcal{L}(\hat{\epsilon}, \boldsymbol{\theta}, \mathbf{g}) = \frac{1}{2\sigma^2} (\mathbf{r} - \hat{\mathbf{E}}\mathbf{P}\check{\mathbf{F}}^H \mathbf{D}\mathbf{W}\mathbf{g})^H (\mathbf{r} - \hat{\mathbf{E}}\mathbf{P}\check{\mathbf{F}}^H \mathbf{D}\mathbf{W}\mathbf{g}) + \frac{1}{2} \boldsymbol{\theta}^T \Phi^{-1} \boldsymbol{\theta}. \quad (19)$$

- *CIR Estimation*: Solve  $\partial \mathcal{L}(\hat{\epsilon}, \boldsymbol{\theta}, \mathbf{g}) / \partial \mathbf{g}^* = \mathbf{0}$ , we obtain

$$\hat{\mathbf{g}} = (4\rho^2)^{-1} \mathbf{W}^H \mathbf{D}^H \check{\mathbf{F}} \mathbf{P}^H \hat{\mathbf{E}}^H \mathbf{r}. \quad (20)$$

- Substituting  $\mathbf{g} = \hat{\mathbf{g}}$  back into  $\mathcal{L}(\hat{\epsilon}, \boldsymbol{\theta}, \mathbf{g})$  produces  $\mathcal{L}(\hat{\epsilon}, \boldsymbol{\theta})$ .

## IV. Near Optimal Estimator: PHN Estimation

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- *PHN Estimation*: Solve  $\partial \mathcal{L}(\hat{\epsilon}, \boldsymbol{\theta}) / \partial \boldsymbol{\theta} = \mathbf{0}$ , we obtain

$$\hat{\boldsymbol{\theta}} = [\text{Re}(\hat{\mathbf{E}}\mathbf{A}\hat{\mathbf{E}}^H) + 4\sigma^2\rho^2\boldsymbol{\Phi}^{-1}]^{-1}\text{Im}(\hat{\mathbf{E}}\mathbf{A}\hat{\mathbf{E}}^H)\mathbf{1}. \quad (21)$$

- Combining the above steps leads to the *Modified Joint CFO/PHN/CIR Estimation (Modified JCPCE)* algorithm with closed-form CFO estimation:

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$$\begin{aligned} \text{Step 1: } & \hat{\epsilon} = \frac{1}{\pi} \angle \mathbf{r}_1^H (\mathbf{R}_1 \boldsymbol{\Phi}_\Delta \mathbf{R}_1^H + 4\sigma^2 \mathbf{I})^{-1} \mathbf{r}_2; \\ & \hat{\mathbf{E}} = \text{diag}([1, e^{j2\pi\hat{\epsilon}/N}, \dots, e^{j2\pi(N-1)\hat{\epsilon}/N}]^T); \\ \text{Step 2: } & \hat{\boldsymbol{\theta}} = [\text{Re}(\hat{\mathbf{E}}\mathbf{A}\hat{\mathbf{E}}^H) + 4\sigma^2\rho^2\boldsymbol{\Phi}^{-1}]^{-1}\text{Im}(\hat{\mathbf{E}}\mathbf{A}\hat{\mathbf{E}}^H)\mathbf{1}; \\ & \hat{\mathbf{P}} = \text{diag}([e^{j\hat{\theta}_0}, \dots, e^{j\hat{\theta}_{N-1}}]^T); \\ \text{Step 3: } & \hat{\mathbf{g}} = (4\rho^2)^{-1} \mathbf{W}^H \mathbf{D}^H \check{\mathbf{F}} \hat{\mathbf{P}}^H \hat{\mathbf{E}}^H \mathbf{r}. \end{aligned}$$


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## V. Complexity Reduction: Special Structure of $\Psi$

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- Letting  $\Psi = \frac{1}{4\sigma^2\rho^2}\Phi$ , the dominant complexity of the previous algorithm is associated with the evaluation of  $[\text{Re}(\mathbf{E}\mathbf{A}\mathbf{E}^H) + \Psi^{-1}]^{-1}$ , which in general has complexity  $\mathcal{O}(N^3)$ .
- Fortunately, complexity reduction is available by noticing the following:
  - For *Wiener PHN*,  $\Psi^{-1}$  is a tridiagonal matrix:

$$\Psi^{-1} = \frac{4\sigma^2\rho^2}{\alpha_\phi^2} \begin{bmatrix} 2 & -1 & & & \mathbf{0} \\ -1 & 2 & -1 & & \\ & \ddots & \ddots & \ddots & \\ & & -1 & 2 & -1 \\ \mathbf{0} & & & -1 & 1 \end{bmatrix}. \quad (22)$$

- For *Gaussian PHN*,  $\Psi$  is a Toeplitz matrix, but can be closely approximated as a circulant matrix  $\tilde{\Psi} = \mathbf{F}\mathbf{\Lambda}\mathbf{F}^H$ . Therefore, its inverse is easy to evaluate,  $\tilde{\Psi}^{-1} = \mathbf{F}\mathbf{\Lambda}^{-1}\mathbf{F}^H$

## V. Complexity Reduction: Conjugate Gradient Method

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- Letting  $\mathbf{q} = \text{Im}(\hat{\mathbf{E}}\mathbf{A}\hat{\mathbf{E}}^H)\mathbf{1}$ , the evaluation of  $[\text{Re}(\mathbf{E}\mathbf{A}\mathbf{E}^H) + \mathbf{\Psi}^{-1}]^{-1}\mathbf{q}$  can be accomplished by the conjugate gradient method as follows:

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Initialization:

$$\boldsymbol{\theta}_0 = \mathbf{0}$$

$$\boldsymbol{\gamma}_0 = [\text{Re}(\mathbf{E}\mathbf{A}\mathbf{E}^H) + \tilde{\mathbf{\Psi}}^{-1}]\boldsymbol{\theta}_0 - \mathbf{q} = -\mathbf{q}$$

$$\boldsymbol{\nu}_0 = -\boldsymbol{\gamma}_0 = \mathbf{q}$$

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For

$$k = 0 : i - 1$$

$$\alpha_k = \boldsymbol{\gamma}_k^H \boldsymbol{\gamma}_k / (\boldsymbol{\nu}_k^H [\text{Re}(\mathbf{E}\mathbf{A}\mathbf{E}^H) + \tilde{\mathbf{\Psi}}^{-1}]\boldsymbol{\nu}_k)$$

$$\boldsymbol{\theta}_{k+1} = \boldsymbol{\theta}_k + \alpha_k \boldsymbol{\nu}_k$$

$$\boldsymbol{\gamma}_{k+1} = \boldsymbol{\gamma}_k + \alpha_k [\text{Re}(\mathbf{E}\mathbf{A}\mathbf{E}^H) + \tilde{\mathbf{\Psi}}^{-1}]\boldsymbol{\nu}_k$$

$$\beta_{k+1} = \frac{\boldsymbol{\gamma}_{k+1}^H \boldsymbol{\gamma}_{k+1}}{\boldsymbol{\gamma}_k^H \boldsymbol{\gamma}_k}$$

$$\boldsymbol{\nu}_{k+1} = -\boldsymbol{\gamma}_{k+1} + \beta_{k+1} \boldsymbol{\nu}_k$$

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End

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## V. Complexity Reduction: Overall Complexity

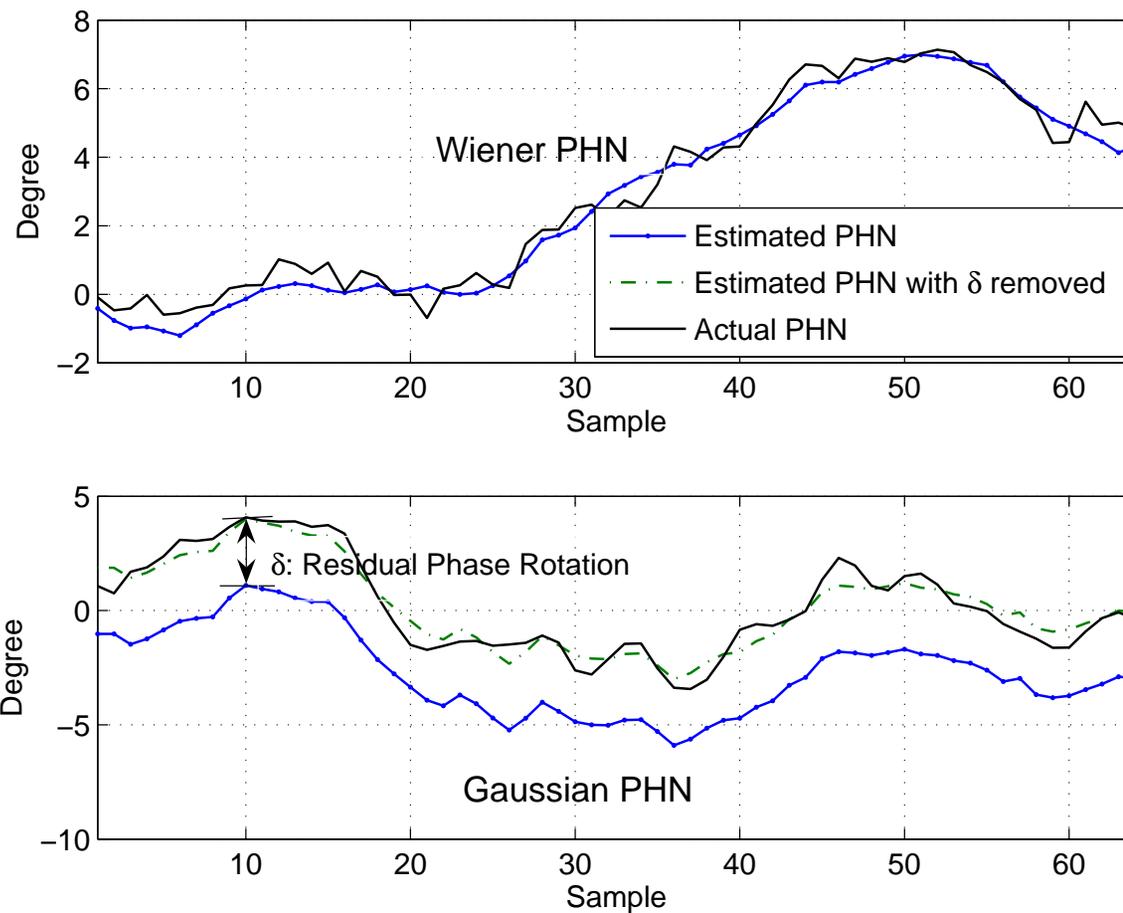
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- By utilizing the special structure of  $\Psi$ , we can reduce the complexity from  $\mathcal{O}(N^3)$  to  $\mathcal{O}(N^2 \log N)$ .
- Using the Conjugate Gradient algorithm, we have control over the number of iteration  $i$ . Exact matrix inversion corresponds to  $i = N$ .
- In practice,  $i \ll N$ . In this case, the complexity becomes  $\mathcal{O}(i \times N \log N)$ .
- Simulations demonstrate that even for  $i = 5$ , no significant performance degradation results.

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## VI. Residual Phase Rotation

- *Unresolvable residual common phase rotation*: A residual phase rotation  $\delta$  cannot be estimated for Gaussian PHN.  $\delta \sim \mathcal{N}(0, \mathbf{1}^T \mathbf{\Phi} \mathbf{1} / N^2)$ .



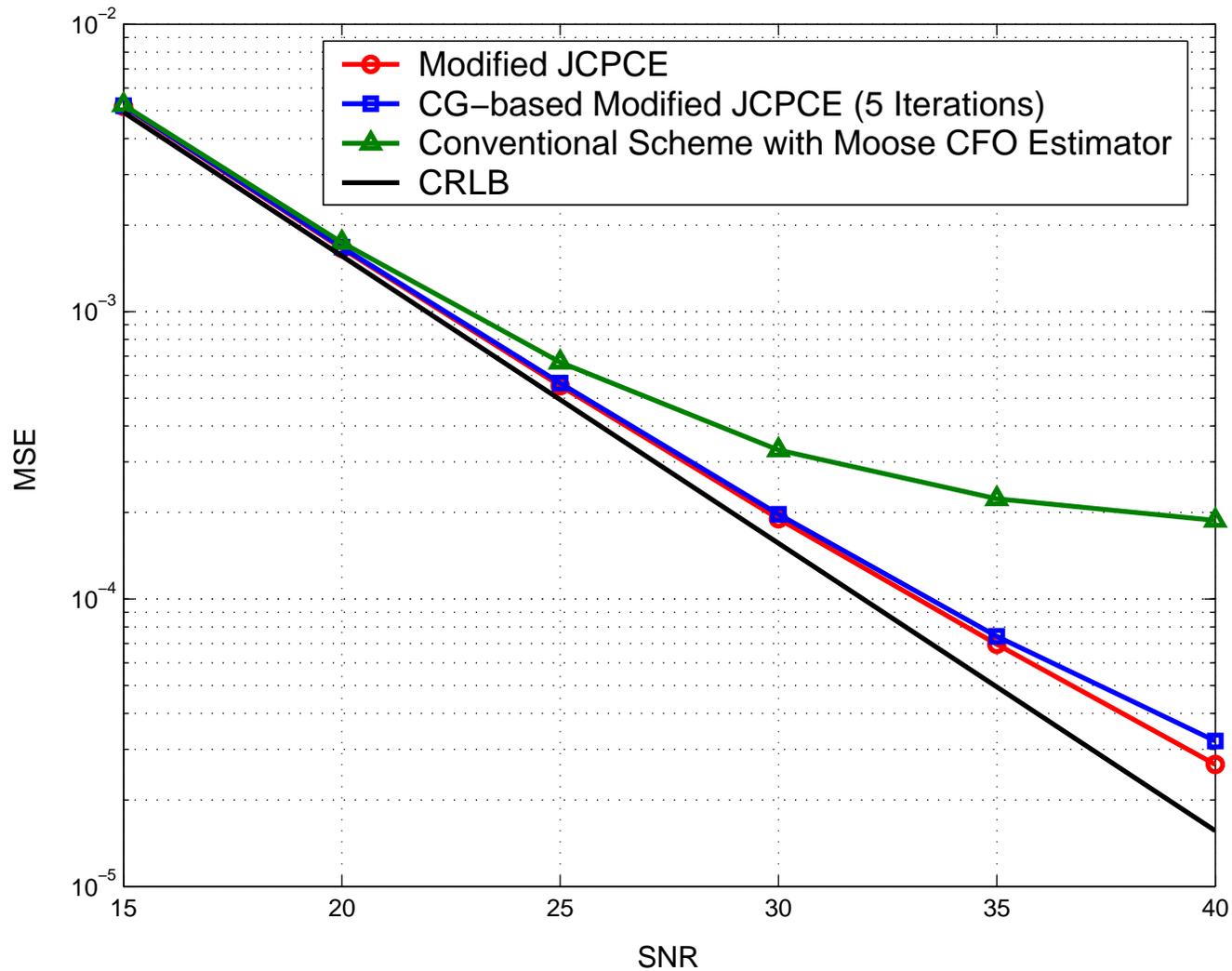
## VI. Simulation Settings

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- The effect of residual phase rotation is that the channel estimate  $\hat{\mathbf{g}}$  is off by a small unknown phase  $\delta$ , which does not introduce ICI.
- We assume that  $\delta$  can be perfectly corrected to facilitate easy assessment of channel estimation mean-squared error (MSE).
- The following system parameters are used in simulations:
  - A Rayleigh multipath fading channel with a delay of  $L = 10$  taps;
  - An OFDM training symbol size of  $N = 64$  subcarriers with each subcarrier modulated in QPSK;
  - Baseband sampling rate  $f_s = 20$  MHz;
  - The Wiener PHN is generated as a random-walk process with incremental PHN of  $\alpha_\phi = 1$  deg.
  - The Gaussian PHN has a standard deviation of  $\theta_{rms} = 4$  deg. It is generated as i.i.d. Gaussian samples passed through a single pole Butterworth filter of 3dB bandwidth  $\Omega_o = 100$  KHz.

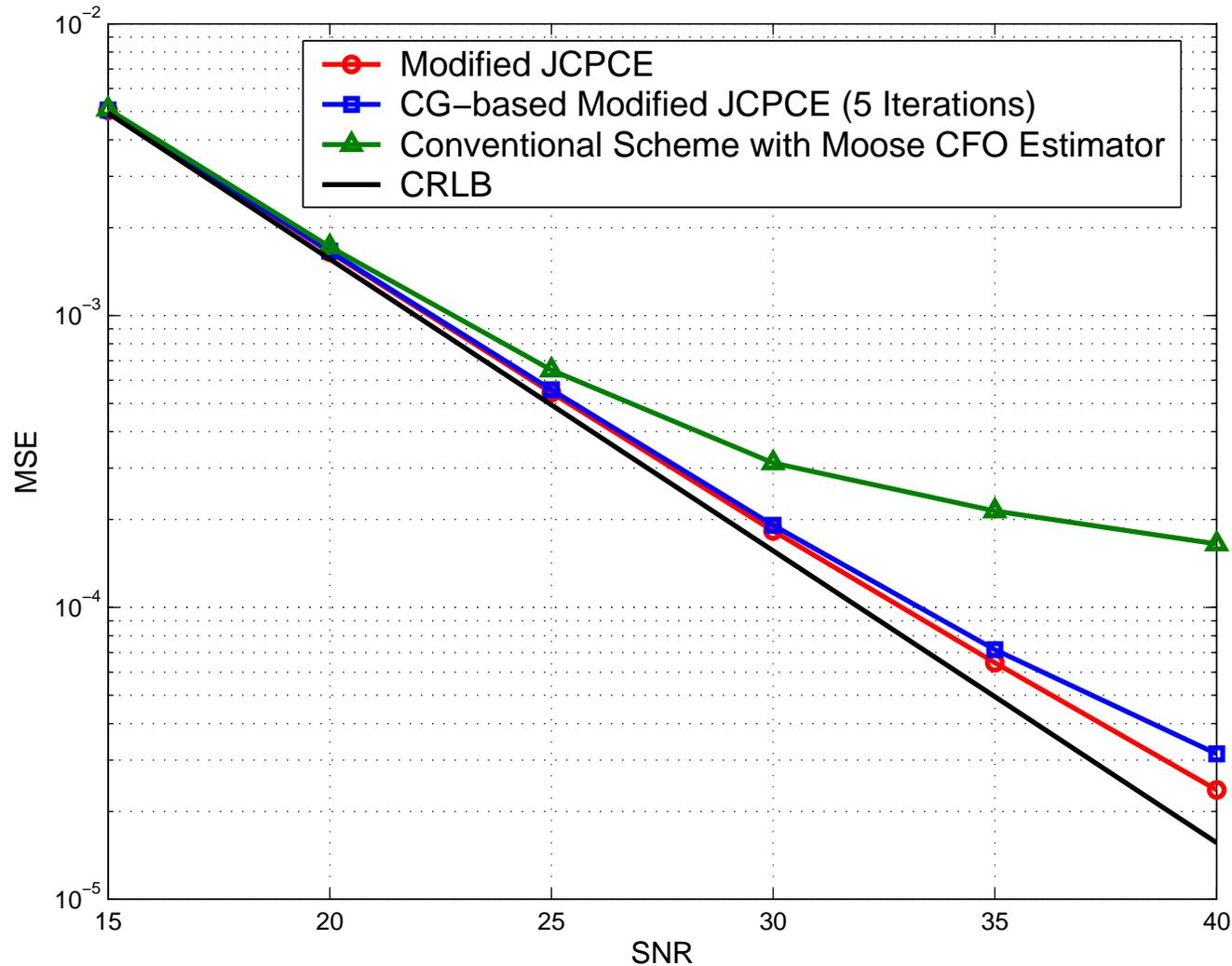
## VI. Simulation Results

- Performance of modified JCPCE in Wiener PHN:



## VI. Simulation Results

- Performance of modified JCPCE in Gaussian PHN:



- *Challenge*: How do we optimally estimate the CIR and CFO of an OFDM channel in the presence of unknown PHN?
- *Solution 1*: In a previous work, we derived a *Joint CFO/PHN/CIR Estimation* (JCPCE) algorithm that optimizes the “complete likelihood function”.
- *Solution 2*: Here we proposed a suboptimal algorithm, called *Modified JCPCE*, to obtain closed-form estimate of CFO, lowering the computational complexity.
- We may reduce the complexity even further using the Conjugate Gradient method.

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**Thank You!**